

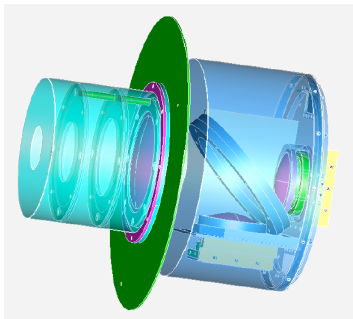
Readout electronics for NIKA

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C. Vescovi

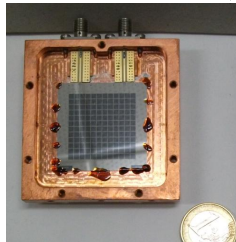
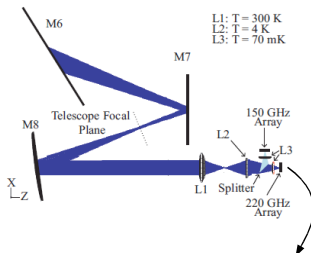
CNRS-IN2P3-LPSC Grenoble

October 3, 2011

NIKA camera



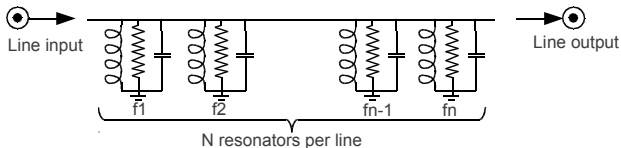
Camera
Field of view 2 arc-min
112 + 72 pixel readout



144 pixels array



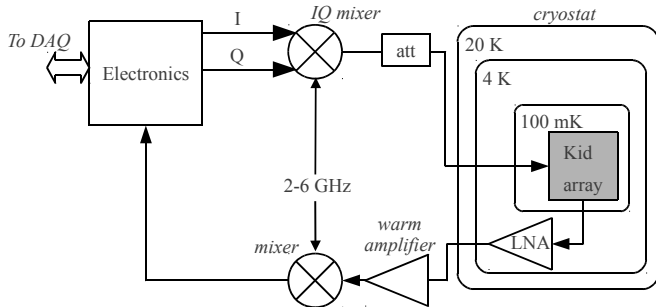
- Bolometer array = line with n high Q resonators
 - Bolometer array built by microlithography
 - **Average** resonator self resonant frequency spacing of 2 MHz.
 - Resonance frequency at several GHz



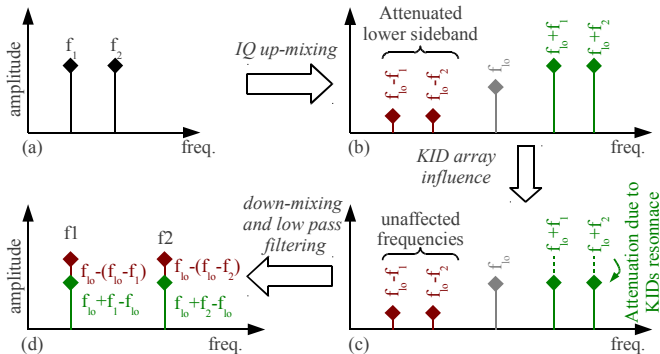
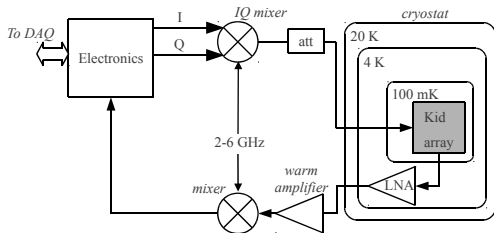
- Detection principle → shift of the resonator self resonant frequency when illuminated.
- **Constraints:**
 - Array at very low temperature (cryostat at 100 mK)
 - Cable feedthrough must be kept to a minimum
 - Work in RF domain
 - Full array readout at several Hertz

Array instrumentation method (1/2)

- 1 Generation of a frequency comb (**frequency multiplexing**):
 - Containing the largest possible number of frequency (\rightarrow Output analog frequency (Shannon))
 - Each frequency adjustable at a kHz resolution.
- 2 Up-mixing of the baseband generated frequency comb
- 3 The signal passes through the array and is eventually modified
- 4 At the array's output, down-mixing to baseband
- 5 Signal processing: determine each tone amplitude and phase

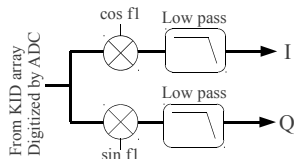


Array instrumentation method (2/2)



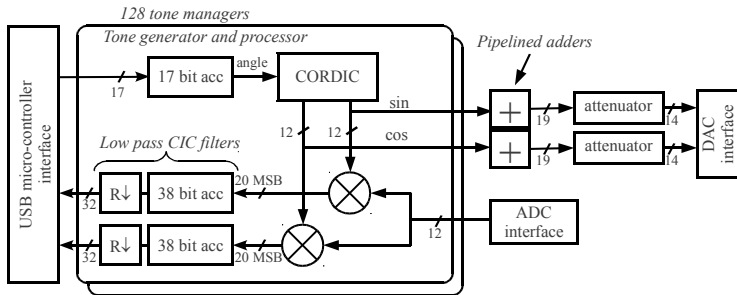
Used methods

- 1 Sine and cosine waves generated by CORDIC (COordinate Rotation Digital Computer)
 - Using only adder/subtractors and a small precomputed arc tangent table
 - Easily algorithm "pipelining"
 - Can reach sine and cosine resolution of $1/2^{n-1}$, where n is the number of +/- stages and atan constants
 - 2 Line output signal analyzed by Digital Down Conversion (DDC)
 - As much processors as spectral rays to analyze
- Principle: signal is shifted to baseband and projected on I, Q axis by multiplying the returning signal by the original sine and cosine wave
 - Low pass filter actually implemented by 1 ms signal averaging (**not moving average → CIC!!**)
 - For $A \times \sin(\omega t + \phi) \Rightarrow |A| = \sqrt{I^2 + Q^2}$ and $\phi = \arctan \frac{Q}{I}$



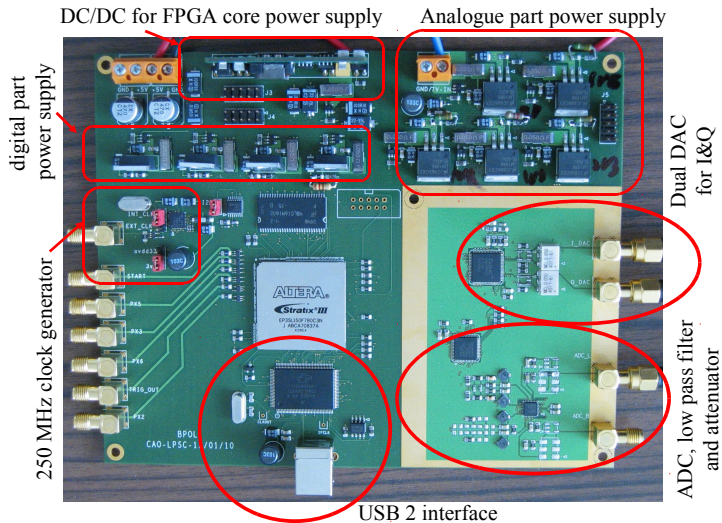


- 128 CORDIC generators
- 128 DDC



First prototype

- $F_{adc}=F_{DAC}=250$ MHz, fanalog max 125 MHz

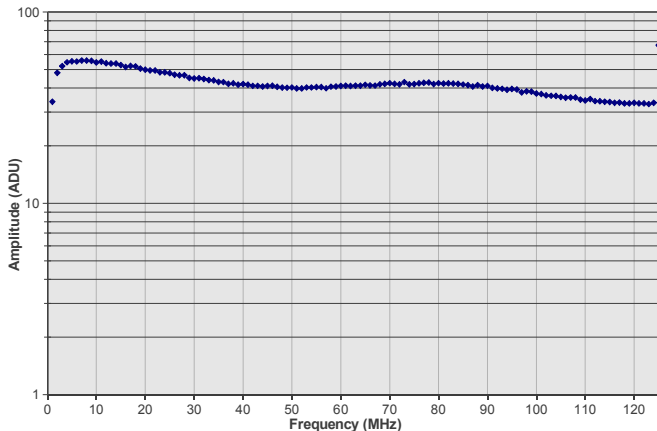


Equipped rack

- Electronic board
- Data server PC (Ubuntu) for Ethernet interfacing
- AC/DC power supplies + fan control



Frequency response

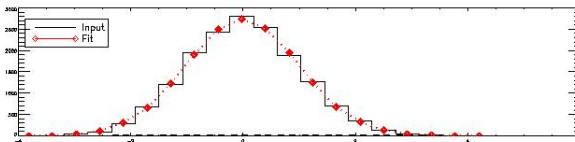
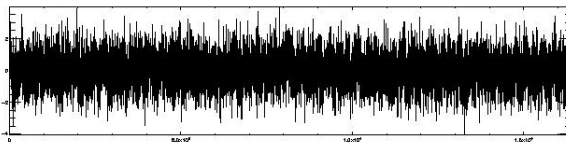
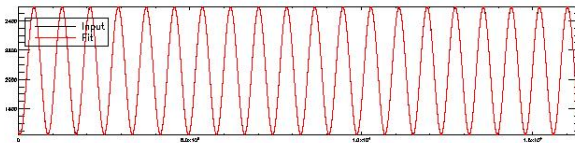


- Flat response over the whole bandwidth (-20 dB over 125 MHz).
- Attenuation mostly due the input Variable Gain Amplifier.

Signal to Noise in loopback mode



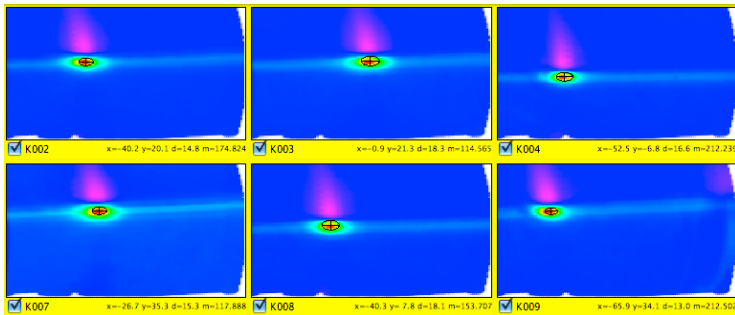
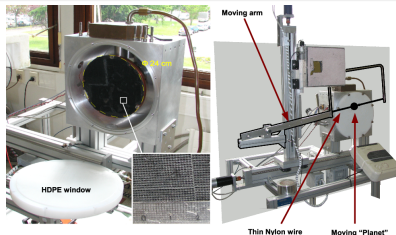
- Sine wave at ~ 300 kHz
- Fitting of a perfect sine wave \rightarrow RMS=1.1
- Electronic chain allows to recover up to ~ 11 effective bit (out of 12)



First images with the sky simulator

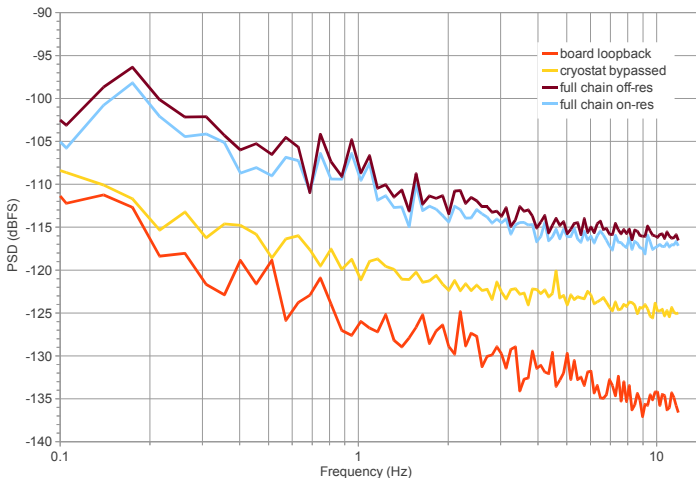


- A source at ambient temperature is placed between a cold plate and the MKID array and moved in 2D



- The “planet” is seen when passing in front of a MKID pixel.

System noise contributor identification

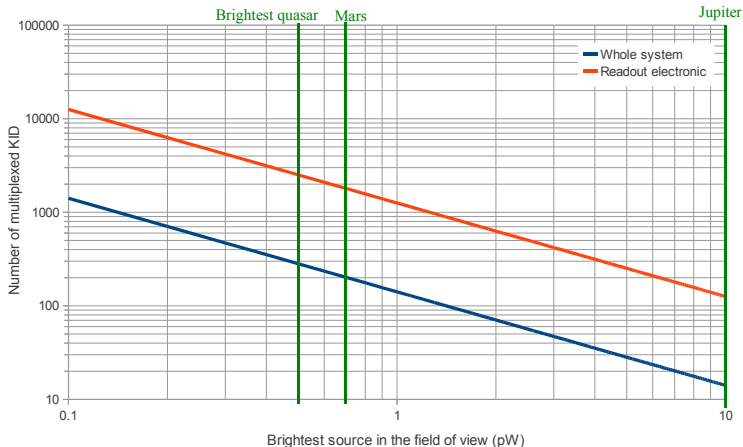


- 1 Noise is not due to the MKIDs themselves (see off/on resonance)
- 2 Noise floor electronic alone at -135 dB
- 3 Noise added by mixing electronics is +10 dB
- 4 Cold amplifier adds 10 dB of noise

Multiplexing limit



- When targeting a constant resolution of $50 \text{ aW}/\sqrt{\text{Hz}}$ at an acquisition rate of 20 Hz with a fixed dynamic range of 115/135 dB, the multiplexing limit is fixed by the brightest spot in the FOV

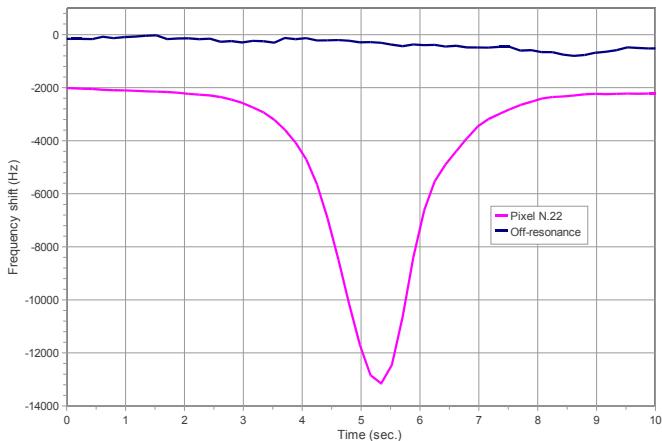


- Mars in FOV \rightarrow 110 MKIDs, Jupiter in FOV \rightarrow 10 MKIDs!!

Crosstalk between MKIDs



- Using sky simulator with cold plate at 50 K
- Highly emissivity ball simulating the planet at 300 K



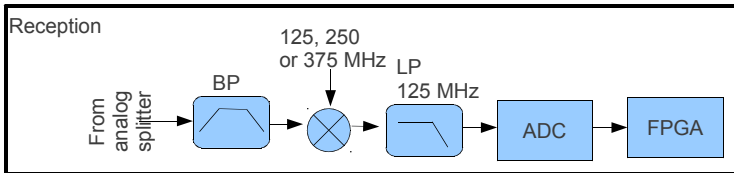
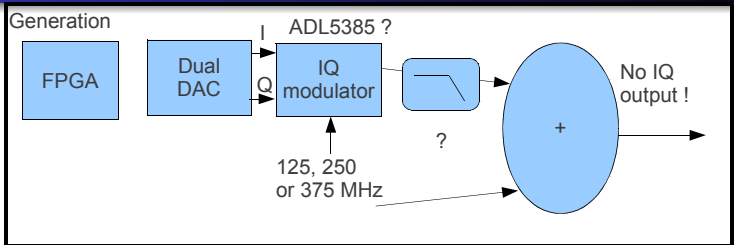
Summary 1st prototype

- Technology and know-how validated for 128 spectral rays fitting in a bandwidth of 125 MHz
 - Validation done with 64 bands separated in average by 2 MHz
- see **2011 JINST 6 P06012**, also available at arXiv:1102.1314

Future

- Equip arrays featuring 2000+4000 pixels
- Design and construct electronic boards able to manage 256 spectral rays over a bandwidth of 500 MHz
 - Limited by trade off between number of spectral rays and multiplexing limit
 - Technical difficulty (computing power/FPGA size, bandwidth, ...)

Obvious (?) "analog" solution



- Idea: split the 500 MHz bandwidth in 4 bands: 0→125, 125→250, 250→375, 375→500 MHz
- 1st drawback: impossible to build directly the comb in IQ
- 2nd drawback: practical impossibility to have highly selective filters
massive crosstalk due to image frequencies!!



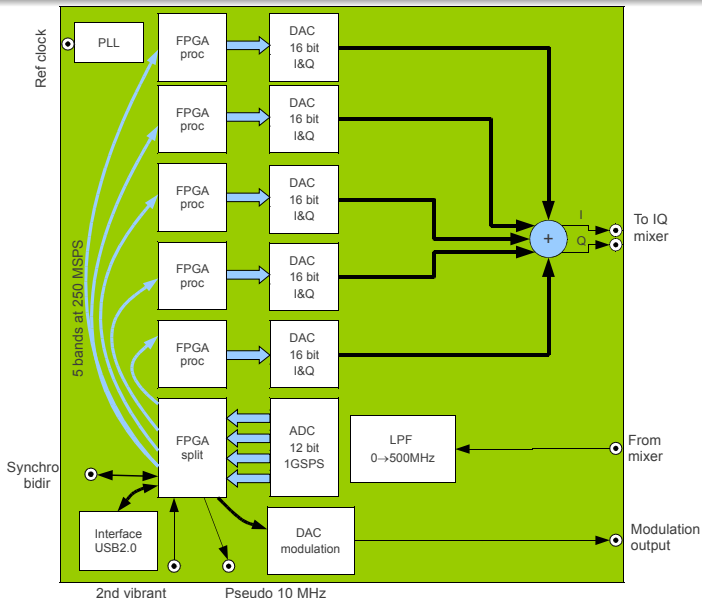
Frequency comb generation

- Nowadays DAC providing analog samples at 1 GSPS while receiving digital inputs at 250 MSPS exist off the shelves
- **Principle:** features digital modulators, interpolators and very steep (high order) filters
- \Rightarrow Use 5 DACs to build 5 100 MHz wide bands to generate the excitation comb

Returning signal digitization and processing

- ADC running at 1 GSPS exist but impossible to manage so many spectral rays in a single FPGA
- Using polyphase filtering and digital demodulation, it is possible to separate the 5 bands and bring them back in baseband for a “regular” processing.
- Becomes almost identical to 5 times the processing developed and implemented in version 1

Board block diagram



Conclusion

- One board will instrument between 256 (2 MHz spacing) and 500 pixels (1 MHz spacing)
- \Rightarrow For 4000 pixels, less than 16 boards required
- Algorithms of the frequency band separator developed and coded in VHDL
- New board available and in debug

